A Second-Order Blind Source Separation Method for Bilinear Mixtures

Lina Jarboui · Yannick Deville · Shahram Hosseini · Rima Guidara · Ahmed Ben Hamida · Leonardo T. Duarte

Received: date / Accepted: date

Abstract In this paper, we are interested in the problem of Blind Source Separation (BSS) using a Second-order Statistics (SOS) method in order to separate autocorrelated and mutually independent sources mixed according to a bilinear (BL) model. In this context, we propose a new approach called *Bilinear Second-order Blind Source Separation* (B-SO-BSS), which is an extension of linear SOS methods, devoted to separate sources present in BL mixtures. These sources, called *extended sources*, include the *actual sources* and their products. We first study the statistical properties of the different *extended sources*, in order to verify the assumption of identifiability when the *actual sources* are zero-mean and when they are not. Then, we present the different steps performed in order to estimate these *actual centred sources* and to extract the actual mixing parameters. The obtained

Yannick Deville

Institut de Recherche en Astrophysique et Planétologie (IRAP), Toulouse University, CNRS-OMP, Toulouse, France. E-mail: yannick.deville@irap.omp.eu

E man. yamnek.aevine@nap.omp.ea

Shahram Hosseini

Institut de Recherche en Astrophysique et Planétologie (IRAP), Toulouse University, CNRS-OMP, Toulouse, France. E-mail: shahram.hosseini@irap.omp.eu

Rima Guidara

Advanced Technologies for Medicine and Signals (ATMS), Sfax University, ENIS, Sfax, Tunisia. E-mail: rima.guidara@gmail.com

Ahmed Ben Hamida

 $\label{eq:advanced} Advanced \ Technologies \ for \ Medicine \ and \ Signals \ (ATMS), \ Sfax \ University, \ ENIS, \ Sfax, \ Tunisia. E-mail: \ Ahmed. Benhamida@enis.rnu.tn$

Leonardo T. Duarte

School of Applied Sciences, University of Campinas (UNICAMP), Limeira, Brazil. E-mail: leonardo.duarte@fca.unicamp.br

Lina Jarboui

Institut de Recherche en Astrophysique et Planétologie (IRAP), Toulouse University, CNRS-OMP, Toulouse, France.

Advanced Technologies for Medicine and Signals (ATMS), Sfax University, ENIS, Sfax, Tunisia. E-mail: Lina.Jarboui@irap.omp.eu

results using artificial mixtures of synthetic and real sources confirm the effectiveness of the new proposed approach.

1 Introduction

Blind Source Separation (BSS) consists in decomposing several observed signals into a set of source signals and their mixing parameters, with almost no prior knowledge about them (Comon and Jutten, 2010; Deville, 2016). Most of the researches dealing with BSS methods suppose that the mixing model is linear, where the observed signals result from linear combinations of the source signals. Nevertheless, for some applications, the linear mixing model is not valid, and must be replaced by a nonlinear one. This nonlinear model provides a better description of the mixing process and the interactions between sources. Due to the complexity of nonlinear models, the nonlinear BSS methods are more complex and remain less studied (Taleb, 2002; Deville and Hosseini, 2009; Hosseini and Deville, 2013). This complexity may be reduced by constraining the structure of the mixing models. Indeed, the addition of simplifying assumptions has allowed the development of exploitable nonlinear models. Among these models, the Linear-Quadratic (LQ) model has drawn significant attention (see e.g. the survey in (Deville and Duarte, 2015)). Research focused on the LQ model has demonstrated the relevance of its use in several applications such as remote sensing (Meganem et al, 2014a,b; Eches and Guillaume, 2014; Jarboui et al, 2014, 2016), analysis of gas sensor array data (Bedoya, 2006; Ando et al, 2015), and scanned document processing (Merrikh-Bayat et al, 2011; Duarte et al, 2011; Almeida and Almeida, 2012; Liu and Wang, 2013). The particularity of the LQ model as compared with the linear one is the presence of the second-order terms. Thereby, considering K observations resulting from an LQ mixture of L sources, the relationship between the observed and the source signals can be characterized by the following equation

$$x_i(n) = \sum_{j=1}^{L} a_j(i)s_j(n) + \sum_{j=1}^{L} \sum_{k=j}^{L} a_{j,k}(i)s_j(n)s_k(n),$$
(1)

where $x_i(n)$ is the i^{th} observed signal at time $n, s_j(n)$ is the j^{th} unknown source signal, $a_j(i)$ is the linear coefficient associated with the j^{th} source signal and the i^{th} observed signal, $a_{j,k}(i)$ is the quadratic mixing coefficient associated with the i^{th} observed signal and resulting from the interaction between the j^{th} and the k^{th} sources. All the actual sources s_j and the pseudo-sources $s_j \times s_k$ are called extended sources, as in (Deville and Duarte, 2015; Meganem et al, 2014b). There is a particular case of the LQ model called the bilinear (BL) model where the squared term coefficients $a_{j,j}(i)$ are null. The model (1) then becomes

$$x_i(n) = \sum_{j=1}^{L} a_j(i)s_j(n) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)s_j(n)s_k(n).$$
(2)

Various methods applicable to the LQ model have been proposed (Deville and Duarte, 2015). While some of them are only devoted to Blind Mixture Identification (BMI), in order to only identify the mixing parameters (Krob and Benidir, 1993; Abed-Meraim et al, 1996), the others are dedicated to BSS, which aims also at estimating the source signals. They include Sparse Component Analysis (SCA) methods (Jarboui et al, 2014; Deville and Hosseini, 2007), which are only applicable to sparse sources, Non-negative Matrix Factorization (NMF) methods (Meganem et al, 2014b; Eches and Guillaume, 2014; Jarboui et al, 2016), which may be used only when sources and mixing coefficients are non-negative, and Independent Component Analysis (ICA) methods (Deville and Hosseini, 2009; Hosseini and Deville, 2013; Castella, 2008), which are based on the assumption that the source signals are statistically independent. Nevertheless, the use of LQ ICA-based methods, either BMI or BSS, is usually constrained by other properties of the sources and/or the mixture. For example, some of them can be used only when the sources are complex-valued and circular (Krob and Benidir, 1993; Abed-Meraim et al, 1996) or binary (Castella, 2008). Others are suited only to determined mixtures (Deville and Hosseini, 2009; Hosseini and Deville, 2013; Almeida and Almeida, 2012), which generally means overlooking a useful part of the available observations. Moreover, most of the existent LQ ICA-based methods are time-consuming.

In this paper, we study the bilinear model in an over-determined configuration. In addition, we suppose that the sources are real-valued, stochastic, auto-correlated, mutually independent and jointly strict-sense stationary signals. Then, we propose a new and fast BSS method, called *Bilinear Second-Order Blind Source Separation* (B-SO-BSS), based on Second-order Statistics (SOS) and the joint diagonalization of correlation matrices of the whitened centred observed signals. Such SOS methods have already been proposed in the framework of linear BSS (Tong et al, 1990; Belouchrani et al, 1997). We first study the correlation between different *extended sources* in section 2, then we present our proposed method developed based on the results of this study in section 3, and in section 4, we eventually provide some simulation results using artificial mixtures of synthetic and real-world sources.

2 Mutual correlation of the extended sources

In this section, by supposing that the *actual sources* s_j are real-valued, stochastic, auto-correlated, mutually independent and jointly strict-sense stationary, we investigate whether all the *extended sources* are mutually uncorrelated in two different cases: when the *actual sources* are zero-mean and when they are not.

2.1 Case of zero-mean *actual sources*

We here detail the study of the correlation between the different *extended sources* when the *actual sources* are zero-mean. The *pseudo-sources* are then zero-mean, as will now be shown. Indeed, these *pseudo-sources* are defined as $s_j(n)s_k(n)$, with $j \neq k$. Their factors $s_j(n)$ and $s_k(n)$ are independent and thus uncorrelated, which yields

$$E\{s_j(n)s_k(n)\} = E\{s_j(n)\}E\{s_k(n)\} = 0$$
(3)

where $E\{.\}$ stands for expectation.

All *extended sources* are thus zero-mean in the considered case. Therefore, their cross-covariance functions, which should be used to measure their correlation, are here equal to their cross-correlation functions. The latter functions are derived hereafter.

2.1.1 Correlation function of $s_i(n)$ and $s_j(n)$

Two different actual sources $s_i(n)$ and $s_j(n)$ being independent and zero-mean, their cross-correlation function is equal to zero:

$$R_{s_i,s_j}(\tau) = E\{s_i(n+\tau)s_j(n)\} = 0.$$
(4)

2.1.2 Correlation function of $s_i(n)$ and $s_j(n) \times s_k(n)$

If $(i \neq j)$ and $(i \neq k)$, since $s_i(n)$, $s_j(n)$ and $s_k(n)$ are independent, we can write

$$R_{s_i,(s_j s_k)}(\tau) = E\{s_i(n+\tau)(s_j(n)s_k(n))\} = E\{s_i(n+\tau)\}E\{s_j(n)\}E\{s_k(n)\} = 0.$$
(5)

If (i = j) or (i = k), the reasoning is similar for the two cases, e.g. considering (i = j), we get

$$R_{s_i,(s_is_k)}(\tau) = E\{s_i(n+\tau)(s_i(n)s_k(n))\} = E\{s_i(n+\tau)s_i(n)\}E\{s_k(n)\} = 0.$$
(6)

2.1.3 Correlation function of $s_i(n) \times s_j(n)$ and $s_k(n) \times s_l(n)$

If $(i \neq k)$, $(i \neq l)$, $(j \neq k)$ and $(j \neq l)$, the independence of $s_i(n)$, $s_j(n)$, $s_k(n)$ and $s_l(n)$ yields

$$R_{(s_i s_j),(s_k s_l)}(\tau) = E\{(s_i(n+\tau)s_j(n+\tau))(s_k(n)s_l(n))\}$$

= $E\{s_i(n+\tau)\}E\{s_j(n+\tau)\}E\{s_k(n)\}E\{s_l(n)\}$
= 0. (7)

If (i = (k or l)) xor (j = (k or l)), the reasoning is similar for all cases, e.g. considering (i = k) and therefore $(j \neq (k \text{ and } l))$, the sources s_i, s_j and s_l are independent so that

$$R_{(s_i s_j),(s_i s_l)}(\tau) = E\{(s_i(n+\tau)s_j(n+\tau))(s_i(n)s_l(n))\}$$

= $E\{s_i(n+\tau)s_i(n)\}E\{s_j(n+\tau)\}E\{s_l(n)\}$
= 0. (8)

Thus, in the case of zero-mean *actual sources*, all the *extended sources* are mutually uncorrelated.

2.2 Case of non-zero-mean actual sources

In this case, (3) shows that the *pseudo-sources* are non-zero-mean, so that the correlation of the *extended sources* must be measured using their cross-covariance functions. Assuming that the *actual sources* are auto-correlated such that the covariance function $C_{s_i,s_i}(\tau) \neq 0$ at a lag τ , we here present two types of extended sources which are mutually correlated.

2.2.1 Covariance function of $s_i(n)$ and $s_j(n) \times s_k(n)$ when (i = j) or (i = k)

The reasoning is similar for the two cases, e.g. considering (i = j), calculating the covariance function yields

$$C_{s_{i},(s_{i}s_{k})}(\tau) = E\{s_{i}(n+\tau)(s_{i}(n)s_{k}(n))\} - E\{s_{i}(n+\tau)\}E\{s_{i}(n)s_{k}(n)\}$$

$$= E\{s_{i}(n+\tau)s_{i}(n)\}E\{s_{k}(n)\} - E\{s_{i}(n+\tau)\}E\{s_{i}(n)\}E\{s_{k}(n)\}$$

$$= \left(E\{s_{i}(n+\tau)s_{i}(n)\} - E\{s_{i}(n+\tau)\}E\{s_{i}(n)\}\right)E\{s_{k}(n)\}$$

$$= C_{s_{i},s_{i}}(\tau)E\{s_{k}(n)\}$$

$$\neq 0$$
(9)

Therefore, $s_i(n)$ and $s_i(n) \times s_k(n)$ are correlated.

2.2.2 Covariance function of $s_i(n) \times s_j(n)$ and $s_k(n) \times s_l(n)$ when (i = (k or l)) xor (j = (k or l))

The reasoning is similar for all cases, e.g. considering (i = k) and therefore $(j \neq (k \text{ and } l))$, the covariance function reads

$$C_{(s_{i}s_{j}),(s_{i}s_{l})}(\tau) = E\{(s_{i}(n+\tau)s_{j}(n+\tau))(s_{i}(n)s_{l}(n))\} - E\{s_{i}(n+\tau)s_{j}(n+\tau)\}E\{s_{i}(n)s_{l}(n)\} = \left(E\{s_{i}(n+\tau)s_{i}(n)\}E\{s_{j}(n+\tau)\}E\{s_{l}(n)\}\right) - \left(E\{s_{i}(n+\tau)\}E\{s_{j}(n+\tau)\}E\{s_{i}(n)\}E\{s_{l}(n)\}\right) = C_{s_{i},s_{i}}(\tau)E\{s_{j}(n+\tau)\}E\{s_{l}(n)\} = C_{s_{i},s_{i}}(\tau)E\{s_{j}(n+\tau)\}E\{s_{l}(n)\}$$

$$\neq 0$$
(10)

Therefore, $s_i(n) \times s_j(n)$ and $s_i(n) \times s_l(n)$ are correlated.

3 Proposed BSS method

In this section, we propose a new BSS method, called *Bilinear Second-Order Blind Source Separation* (B-SO-BSS), first for zero-mean and then for non zero-mean *actual sources*.

3.1 Case of zero-mean *actual sources*

The bilinear mixing model (2) can be written in the following matrix form

$$\mathbf{x}(n) = \mathbf{As}(n),\tag{11}$$

where $\mathbf{x}(n) = [x_1(n), \dots, x_K(n)]^T$ is the vector of K observed signals at time n, $\mathbf{s}(n) = [s_1(n), \dots, s_L(n), s_1(n)s_2(n), \dots, s_{L-1}(n)s_L(n)]^T$ is the vector of all the *extended sources* at time n, and the mixing matrix **A**, which contains both linear and quadratic mixing parameters, reads

$$\mathbf{A} = \begin{pmatrix} a_1(1) \cdots a_L(1) & a_{1,2}(1) \cdots a_{L-1,L}(1) \\ \vdots & \ddots & \vdots & \vdots & \ddots & \vdots \\ a_1(K) \cdots a_L(K) & a_{1,2}(K) \cdots & a_{L-1,L}(K) \end{pmatrix}.$$
 (12)

Then, the bilinear mixture can be considered as a linear mixture of the L(L+1)/2extended sources. In the following, we assume that $K \ge L(L+1)/2$ so that this reformulated linear mixture is not under-determined. As shown in Section 2.1, all the extended sources are mutually uncorrelated in the case considered here. If they are also auto-correlated with different auto-correlation functions, the source separation may be achieved by jointly diagonalizing the correlation matrices of the whitened centred observations at different lags as will be detailed in Section 3.4.

$3.2\ {\rm Case}$ of non-zero-mean $actual\ sources$

As shown in Section 2.2, in this case some *extended sources* are mutually correlated. Here, we show how the original bilinear mixing model may be used to derive a new mixing model with new mutually uncorrelated *extended sources*. The mean $E\{s_j(n)\}$ of the *actual source* $s_j(n)$ does not depend on the considered time n, since the *actual sources* are assumed to be strict-sense stationary. The expectation of s_j will be denoted by \bar{s}_j hereafter. The centred version of $s_j(n)$ is thus $\tilde{s}_j(n) = s_j(n) - \bar{s}_j$. The bilinear model (2) can then be written as

$$x_{i}(n) = \sum_{j=1}^{L} a_{j}(i)(\tilde{s}_{j}(n) + \bar{s}_{j}) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)(\tilde{s}_{j}(n) + \bar{s}_{j})(\tilde{s}_{k}(n) + \bar{s}_{k})$$

$$= \sum_{j=1}^{L} a_{j}(i)\tilde{s}_{j}(n) + \sum_{j=1}^{L} a_{j}(i)\bar{s}_{j} + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\tilde{s}_{j}(n)\tilde{s}_{k}(n)$$

$$+ \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)(\bar{s}_{k}\tilde{s}_{j}(n) + \bar{s}_{j}\tilde{s}_{k}(n)) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_{j}\bar{s}_{k}.$$
(13)

The fourth term on the right hand side of (13), denoted as F in the following, can be rewritten as

$$F = \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_k \tilde{s}_j(n) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_j \tilde{s}_k(n)$$

=
$$\sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_k \tilde{s}_j(n) + \sum_{k=1}^{L-1} \sum_{j=k+1}^{L} a_{k,j}(i)\bar{s}_k \tilde{s}_j(n), \qquad (14)$$

where the last term above is obtained just by inverting the roles of symbols j and k.

Then, we introduce the coefficients $a_{j,k}(i)$ with j > k, defined with respect to the actual coefficients of (2), as $a_{j,k}(i) = a_{k,j}(i)$. This yields

$$F = \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i) \bar{s}_k \tilde{s}_j(n) + \sum_{k=1}^{L-1} \sum_{j=k+1}^{L} a_{j,k}(i) \bar{s}_k \tilde{s}_j(n)$$
$$= \left(\sum_{j=1}^{L-1} \sum_{k=j+1}^{L} \bigcup_{k=1}^{L-1} \sum_{j=k+1}^{L} \right) a_{j,k}(i) \bar{s}_k \tilde{s}_j(n).$$
(15)

The above sum contains all possible combinations of $j \in [1, L], k \in [1, L]$ such that $j \neq k$. It can then be rewritten as

$$F = \sum_{j=1}^{L} \sum_{k=1, k \neq j}^{L} a_{j,k}(i) \bar{s}_k \tilde{s}_j(n).$$
(16)

Replacing (16) in (13) leads to

$$x_{i}(n) = \sum_{j=1}^{L} \left(a_{j}(i) + \sum_{k=1, k \neq j}^{L} a_{j,k}(i)\bar{s}_{k} \right) \tilde{s}_{j}(n) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\tilde{s}_{j}(n) \tilde{s}_{k}(n) + \sum_{j=1}^{L} a_{j}(i)\bar{s}_{j} + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_{j}\bar{s}_{k},$$
(17)

which yields

$$x_{i}(n) = \sum_{j=1}^{L} \widetilde{a}_{j}(i)\widetilde{s}_{j}(n) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\widetilde{s}_{j}(n)\widetilde{s}_{k}(n) + C_{i},$$
(18)

where $\tilde{a}_{j}(i)$ are the linear coefficients of the new model which are defined as

$$\tilde{a}_{j}(i) = a_{j}(i) + \sum_{k=1, k \neq j}^{L} a_{j,k}(i)\bar{s}_{k},$$
(19)

and C_i is a constant defined as

$$C_i = \sum_{j=1}^{L} a_j(i)\bar{s}_j + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_j\bar{s}_k.$$
 (20)

Since the actual centred sources $\tilde{s}_j(n)$ and $\tilde{s}_k(n)$ are zero-mean and independent, from (18) the mean of the observed value $x_i(n)$ is equal to $\bar{x}_i = C_i$. Thus, its centred version can be written as follows:

$$\widetilde{x}_{i}(n) = x_{i}(n) - \overline{x}_{i}$$

$$= \sum_{j=1}^{L} \widetilde{a}_{j}(i)\widetilde{s}_{j}(n) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\widetilde{s}_{j}(n)\widetilde{s}_{k}(n).$$
(21)

As can be seen, the centred observations form a new bilinear mixture of the *actual* centred sources, although the mixing parameters of the linear part in this new model are not the same as those in the original mixture. According to the results provided in Section 2.1, the new extended sources $\tilde{s}_j(n)$ and $\tilde{s}_j(n)\tilde{s}_k(n)$ are all mutually uncorrelated. We can then rewrite this new bilinear model in the matrix form (11) just by replacing **s** and **x** by \tilde{s} and \tilde{x} , and the parameters $a_j(i)$ by $\tilde{a}_j(i)$ in the expression (12) of the matrix **A** to obtain the matrix \tilde{A} .

The approach that we developed at this stage therefore yields a modified set of observations, namely the centred observations $\tilde{x}_i(n)$, which form a determined (or overdetermined) linear mixture of a modified set of mutually uncorrelated sources signals, namely the *extended centred sources* related to the *actual centred sources*. Moreover, we hereafter consider the case when these modified source signals are auto-correlated with different auto-correlation functions. With respect to these modified observations and source signals, the configuration that we thus derived meets the same main assumptions as those which have previously been used in the literature, for plain linear mixtures, to derive second-order BSS methods, such as the Algorithm for Multiple Unknown Signals Extraction (AMUSE) (Tong et al, 1990) or its improved version, that is the Second-Order Blind Identification (SOBI) method (Belouchrani et al, 1997). This then allows us to derive extended versions of the above standard methods, which were initially intended for linear mixtures, in order to process our configuration based on bilinear mixtures. In particular, we hereafter propose an extension of SOBI.

3.3 Identifiability condition

A necessary step in the proposed method is to check the mixture identifiability condition. Similarly to the SOBI method, the proposed method uses several correlation matrices of the whitened centred observations for a fixed set of different non-zero lags $\tau_i \in {\tau_1, ..., \tau_m}$. For a lag τ_i , the correlation matrix of the L(L+1)/2 whitened centred observations $\mathbf{z}(n) = \mathbf{W} \widetilde{\mathbf{x}}(n)$ (where \mathbf{W} is a whitening matrix) is given by:

$$R_{\mathbf{z}}(\tau_i) = \mathbf{U}R_{\widetilde{\mathbf{s}}}(\tau_i)\mathbf{U}^T \tag{22}$$

where **U** denotes an orthogonal matrix, T stands for transposition, and $R_{\tilde{s}}(\tau_i)$ denotes the correlation matrix of the *extended centred sources* associated with the lag τ_i , which is a diagonal matrix since the *extended centred sources* are mutually uncorrelated.

Let us consider the following theorem:

-

Theorem: Let $\tau_i = {\tau_1, \dots, \tau_m}$ be *m* non-zero lags, **V** be an orthogonal matrix, such that:

$$\forall 1 \le i \le m \quad \mathbf{V}^T R_{\mathbf{z}}(\tau_i) \mathbf{V} = diag[d_1(i), \cdots, d_{L(L+1)/2}(i)]$$
(23)

$$\forall 1 \le j \ne k \le L(L+1)/2, \quad \exists i, 1 \le i \le m \quad d_j(i) \ne d_k(i). \tag{24}$$

Then, **U** and **V** are *essentially* equal, i.e. they are equal up to a multiplication by a matrix **P**, such that $\mathbf{U} = \mathbf{VP}$, where **P** has one nonzero entry in each row and column, whose value is equal to ± 1 .

This theorem provides a uniqueness condition for the matrix **U** and consequently the mixing matrix $\widetilde{\mathbf{A}}$. Note that the mixing matrix cannot be identified when the *extended centred sources* have identical normalized spectra. But, if they have different normalized spectra, it is possible to find a set of lags τ_i satisfying the theorem condition. More details are provided in (Belouchrani et al, 1997).

It should in particular be noted that if one of the *actual centred sources* $\tilde{s}_i(n)$ is temporally uncorrelated, then all the *pseudo-sources* related to it, i.e. $\tilde{s}_i(n)\tilde{s}_j(n)$ with $i \neq j$ are temporally uncorrelated too, so that all these *extended centred sources* have identical (constant) normalized spectra. Thus, a necessary condition for identifiability is that all the *actual centred sources* must be autocorrelated¹.

3.4 Proposed algorithm

Our proposed algorithm (B-SO-BSS), which provides estimates of centred *actual* sources up to a permutation and scale factors, is summarized in Algorithm 1.

Algorithm 1 : B-SO-BSS

- Estimate the zero-lag correlation matrix $R_{\tilde{\mathbf{x}}}(0)$ of the centred observed signals $\tilde{\mathbf{x}}(n)$.

- Calculate the whitening matrix $\mathbf{W} = \mathbf{D}^{-1/2} \mathbf{E}^T$ where \mathbf{D} and \mathbf{E} are, respectively, the matrices containing the eigenvalues on its diagonal and unit-norm eigenvectors of the estimate of $R_{\tilde{\mathbf{x}}}(0)$.

- Whiten the centred observed signals $\widetilde{\mathbf{x}}(n)$: $\mathbf{z}(n) = \mathbf{W}\widetilde{\mathbf{x}}(n)$.

- Estimate the correlation matrices $R_{\mathbf{z}}(\tau_i)$ of \mathbf{z} , where $\tau_i = \{\tau_1, \cdots, \tau_m\}$ are m chosen lags.

- Perform the joint diagonalization of the estimates of matrices $R_{\mathbf{z}}(\tau_i)$ to provide an estimate $\hat{\mathbf{U}}$ of an orthogonal matrix \mathbf{U} so that $R_{\mathbf{z}}(\tau_i) = \mathbf{U}R_{\mathbf{\tilde{s}}}(\tau_i)\mathbf{U}^T$ where $\tau_i = \{\tau_1, \cdots, \tau_m\}$ (Belouchrani et al, 1997).

- Calculate $\hat{\mathbf{s}}(n) = \hat{\mathbf{U}}^T \mathbf{z}(n)$ which provides an estimate of the *extended centred* source vector $\tilde{\mathbf{s}}(n)$, up to permutation and scale indeterminacies.

- Calculate $\hat{\mathbf{A}} = \mathbf{W}^{\dagger} \hat{\mathbf{U}}$ (where \dagger stands for pseudo-inverse) which provides an estimate of the matrix $\widetilde{\mathbf{A}}$, up to permutation and scale indeterminacies.

- Identify the L estimated *actual centred sources* among all the estimated *extended centred sources* according to Section 3.5.

In the particular case in which the *actual sources* are zero-mean, the same algo-

 $^{^1\,}$ Note that in the linear SOBI, at most one source may be temporally uncorrelated.

rithm may be used just by choosing $\tilde{\mathbf{s}} = \mathbf{s}$, $\tilde{\mathbf{A}} = \mathbf{A}$, and $\tilde{\mathbf{x}} = \mathbf{x}$. The last step of the proposed algorithm, which consists in identifying the estimated *actual centred sources* among all the estimated *extended centred sources*, is detailed below.

3.5 Identifying the estimated actual centred sources

The first steps of the proposed method yield a set of signals $\hat{\mathbf{s}}(n)$ composed of estimates of the L(L+1)/2 unordered extended centred sources, up to a permutation and scale factors. Thus, if e.g. $\hat{s}_j(n)$ and $\hat{s}_k(n)$ correspond to two centred *actual* sources and $\hat{s}_i(n)$ corresponds to their product, then $\hat{s}_i(n)$ must ideally be proportional to $\hat{s}_j(n) \times \hat{s}_k(n)$. As a result, the absolute value of the correlation coefficient between $\hat{s}_i(n)$ and $\hat{s}_j(n) \times \hat{s}_k(n)$ must be close to one. Thus, by computing this correlation coefficient for all the possible triplets $\{i, j, k\}, i \neq j \neq k$, we can identify the estimated *actual centred sources* among all the estimated *extended centred sources*².

3.6 Estimation of actual mixing coefficients

Many BSS applications only aim at estimating the *actual source waveforms*, which are provided by our Algorithm 1. In some applications (like in hyperspectral image unmixing), however, it is also needed to estimate the mixing parameters. In the case of non-zero-mean *actual sources*, our algorithm provides an estimate of matrix $\tilde{\mathbf{A}}$, and not \mathbf{A} (up to a permutation and a diagonal matrix). As mentioned in Section 3.2, matrix $\tilde{\mathbf{A}}$ consists of:

- columns containing quadratic coefficients $a_{j,k}(i)$, like in matrix A,
- columns containing modified linear coefficients $\tilde{a}_j(i)$, which are different from the actual coefficients $a_j(i)$ included in matrix **A**, and are defined by (19).

In the following, we propose a method to recover an estimate of the actual matrix \mathbf{A} (up to classical indeterminacies) from the estimate of matrix $\widetilde{\mathbf{A}}$ provided by Algorithm 1.

From (19), we have

$$a_j(i) = \tilde{a}_j(i) - \sum_{k=1, k \neq j}^L a_{j,k}(i)\bar{s}_k.$$
 (25)

² In the special case of L = 2 actual centred sources, it is also possible to identify the estimated actual centred sources among 3 estimated extended centred sources using a criterion measuring statistical independence, like mutual information. Actually, we know that \tilde{s}_1 and \tilde{s}_2 are mutually independent while $\tilde{s}_1 \times \tilde{s}_2$ is not independent from \tilde{s}_1 and \tilde{s}_2 .

Inserting (25) in (20) yields

$$C_{i} = \bar{x}_{i}$$

$$= \sum_{j=1}^{L} [\tilde{a}_{j}(i) - \sum_{k=1, k \neq j}^{L} a_{j,k}(i)\bar{s}_{k}]\bar{s}_{j} + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_{j}\bar{s}_{k}$$

$$= \sum_{j=1}^{L} \tilde{a}_{j}(i)\bar{s}_{j} - \sum_{j=1}^{L} \sum_{k=1, k \neq j}^{L} a_{j,k}(i)\bar{s}_{k}\bar{s}_{j} + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_{j}\bar{s}_{k}$$

$$= \sum_{j=1}^{L} \tilde{a}_{j}(i)\bar{s}_{j} - 2\sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_{j}\bar{s}_{k} + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_{j}\bar{s}_{k}$$

$$= \sum_{j=1}^{L} \tilde{a}_{j}(i)\bar{s}_{j} - \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)\bar{s}_{j}\bar{s}_{k}$$

$$= \sum_{j=1}^{L} \tilde{a}_{j}(i)(\bar{s}_{j}) + \sum_{j=1}^{L-1} \sum_{k=j+1}^{L} a_{j,k}(i)(\bar{s}_{j}\bar{s}_{k}),$$
(26)

where d_j and $d_{j,k}$ are unknown arbitrary scale factors, up to which the *extended* centred sources and the columns of matrix $\widetilde{\mathbf{A}}$ have been estimated by Algorithm 1. The above result can be written in the following matrix form

$$\mathbf{c} = \widetilde{\mathbf{A}}_1 \mathbf{e},\tag{27}$$

where $\mathbf{c} = [C_1, \dots, C_K]^T$, $\mathbf{e} = [d_1 \bar{s}_1, \dots, d_L \bar{s}_L, (-d_{1,2} \bar{s}_1 \bar{s}_2), \dots, (-d_{L-1,L} \bar{s}_{L-1} \bar{s}_L)]^T$, and $\widetilde{\mathbf{A}}_1$ is the result of dividing the columns of matrix $\widetilde{\mathbf{A}}$ by unknown scale factors d_j and $d_{j,k}$. In other words, $\widetilde{\mathbf{A}}_1$ is the matrix $\widehat{\mathbf{A}}$ provided by Algorithm 1 up to estimation errors³.

Note that ${\bf c}$ can easily be obtained by estimating the means of observations. As a result, ${\bf e}$ can be obtained using

$$\mathbf{e} = \widetilde{\mathbf{A}}_{1}^{\dagger} \mathbf{c}, \qquad (28)$$

where *†* stands for pseudo-inverse.

Furthermore, (25) can be rewritten as

$$a_{j}(i)\bar{s}_{j} = \frac{\tilde{a}_{j}(i)}{d_{j}}(d_{j}\bar{s}_{j}) + \sum_{k=1,k\neq j}^{L} \frac{a_{j,k}(i)}{d_{j,k}}(-d_{j,k}\bar{s}_{j}\bar{s}_{k}).$$
(29)

As mentioned above, $\frac{\tilde{a}_j(i)}{d_j}$ and $\frac{a_{j,k}(i)}{d_{j,k}}$ ($\forall i = 1, \dots, K$) correspond to the columns of $\tilde{\mathbf{A}}_1$, estimated by Algorithm 1, and $(d_j \bar{s}_j)$ and $(-d_{j,k} \bar{s}_j \bar{s}_k)$ are the entries of \mathbf{e} ,

³ In fact, the columns of the matrix estimated by Algorithm 1 are also permuted. However, using the method explained in Section 3.5, we can identify the columns containing linear coefficients and the columns containing quadratic ones. It is then possible to arrange the columns of the estimated matrix so that the first L ones correspond to the linear part (their order is not really important) and the other ones are matched correctly to these first L columns.

estimated using (28). Consequently, $a_j(i) \forall i = 1, \dots, K$ can be estimated up to unknown factors \bar{s}_j according to (29). In other words, this approach allows one to estimate the columns of the actual matrix **A** containing the linear coefficients up to scale factors. Note that the columns of this matrix containing the quadratic coefficients are directly provided by Algorithm 1 (up to scale factors too).

4 Simulation results

In this section, we present and discuss the results obtained by the proposed B-SO-BSS method presented in Algorithm 1 to unmix the bilinear mixtures, with adding the step described in Section 3.6 to estimate the actual mixing coefficients. Herein, we just present the results obtained when the sources are non-zero-mean since we found nearly the same performance for both zero-mean and non-zero-mean cases. In our simulations, the processed data are non-negative, and hence it is possible to compare the obtained results to those obtained by the NMF-Grd-LQ algorithm presented in (Meganem et al, 2014b) which is an NMF-based method adapted to LQ mixtures, exploiting the non-negativity of data involved in mixtures. Note that the physical constraints of the NMF-Grd-LQ algorithm, originally adapted to remote sensing applications, *i.e* the sum of the linear coefficients equal to 1 and the quadratic mixing coefficients lower than 0.5, have been omitted in our simulations and the NMF-Grd-LQ method has been modified accordingly.

4.1 Performance criteria

In order to evaluate the performance of the methods, we calculate the Signal-to-Interference Ratio (SIR) and the Normalized Mean Square Error (NMSE) related to each actual centred source according to the following equations

$$SIR_{s_i} = 10 \log_{10} \frac{\sum_{n=1}^{N} \tilde{s}_i(n)^2}{\sum_{n=1}^{N} (\tilde{s}_i(n) - \hat{s}_i(n))^2}$$
(30)

$$NMSE_{s_i} = \frac{\sum_{n=1}^{N} (\tilde{s}_i(n) - \hat{s}_i(n))^2}{\sum_{n=1}^{N} \tilde{s}_i(n)^2}$$
(31)

where N represents the number of available samples for each signal and the notation '^' refers to the estimated values after removing the permutation and scale factor indeterminacies. In the same way, we calculate SIR_a and $NMSE_a$ related to all the mixing parameters, a_j and $a_{j,k}$, estimated according to Section 3.6.

4.2 Tests

We performed the following four experiments:

Experiment 1: we considered artificial mixtures of synthetic sources. The mixing parameters $a_j(i)$ and $a_{j,k}(i)$ were generated randomly with values uniformly distributed between 0 and 1. The generation of two sources was realized as follows: At first, we generated two independent and identically distributed (i.i.d.)

signals $e_1(n)$ and $e_2(n)$ uniformly distributed over [0, 1], then we filtered them by two first-order auto-regressive filters in order to obtain two auto-correlated source signals according to the model $s_i(n) = e_i(n) + \rho_i s_i(n-1)$. The chosen parameters were $\rho_1 = 0.7$ and $\rho_2 = 0.5$. The tests were repeated using different source sample numbers N: 10000, 1000, and 100. Finally, three observed signals were generated using the BL model (2).

Experiment 2: we generated artificial mixtures of two real-world sources. These sources, shown in Fig. 1 and described in (Duarte et al, 2014), correspond to the activities (which can be seen as effective ionic concentrations) of ions Na⁺ and K⁺ measured for 41 samples. As in Experiment 1, the mixing parameters $a_j(i)$ and $a_{j,k}(i)$ were generated with random values uniformly distributed between 0 and 1. Thereafter, we generated three observed signals using the BL model (2), even if the mixture model of the concentrations of the chemical species is usually approximated by a linear-quadratic model.



Fig. 1 Activities of Na+ and K+ ions.

Experiment 3: The third experiment aims at evaluating the robustness of our method to noise. The mixtures were, first, generated in the same way as in Experiment 1, with N = 10000. A zero-mean Gaussian i.i.d noise was then added to the observed signals in order to obtain a noisy setting. The SNR (*Signal to Noise Ratio*) values are varied from 60 dB down to 30 dB.

Experiment 4: With the same strategy concerning noise effect adopted in Experiment 3, we added a zero-mean Gaussian i.i.d. noise to the observed signals generated in Experiment 2. As in Experiment 3, the SNR values are varied from 60 dB down to 30 dB.

4.3 Results

In order to separate the mixed sources and to estimate the mixing parameters, we applied the steps described in Algorithm 1 and Section 3.6 in two different configurations: using only one lag ($\tau_i = 1$), and using 4 lags ($\tau_i = \{1, 2, 3, 4\}$).

In the first two experiments, we performed 100 Monte Carlo simulations for our method and NMF-Grd-LQ. At each simulation, we, randomly, modified the source signals and the mixing parameters in the case of Experiment 1, and only the mixing parameters in the case of Experiment 2. The mean of SIR and NMSE of the sources and all the mixing parameters, and the CPU time ⁴, averaged over all 100 simulations realized in Experiment 1 and Experiment 2, are shown in Table 1.

Algorithms	Number of samples	Synthetic sources 100 samples	Synthetic sources 1000 samples	Synthetic sources 10000 samples	Chemical sources 41 samples
$\boxed{ \begin{array}{c} \textbf{B-SO-BSS} \\ \tau_i = 1 \end{array} }$	$SIR_s(dB)$	9.06	20.18	30.15	13.68
	$NMSE_s$	0.33	0.022	0.0022	0.072
	$SIR_a(dB)$	8.71	11.25	19.38	17.58
	$NMSE_a$	0.36	0.33	0.08	0.31
	CPU Time (sec)	$6.4 \ e^{-4}$	$8.8 e^{-4}$	$2.1 \ e^{-3}$	$3.9 \ e^{-4}$
B-SO-BSS $\tau_i = 1, 2, 3, 4$	$SIR_s(dB)$	8.05	19.54	29.80	15.99
	$NMSE_s$	0.36	0.028	0.0024	0.068
	$SIR_a(dB)$	8.73	11.15	18.89	18.52
	$NMSE_a$	0.39	0.32	0.08	0.31
	CPU Time (sec)	$2.4 \ e^{-3}$	$2.8 \ e^{-3}$	$8.7 \ e^{-3}$	$8.5 \ e^{-4}$
NMF-Grd- LQ	$SIR_s(dB)$	7.65	12.27	14.87	7.48
	NMSEs	0.37	0.069	0.06	0.51
	$SIR_a(dB)$	7.21	7.45	11.09	6.80
	NMSE _a	0.64	0.70	0.18	0.41
	CPU Time (sec)	4.94	19.87	127.77	3.11

Table 1 Simulation results using our algorithm (B-SO-BSS) with one and four lags, and using NMF-Grd-LQ algorithm. SIR_S and $NMSE_S$ refer to sources, while SIR_a and $NMSE_a$ refer to mixing parameters.

As can be seen, our proposed method leads to the best results. In terms of runtime, the proposed method is much faster than NMF-Grd-LQ.

Moreover, to evaluate the quality of the source estimation, comparisons are then

 $^{^4\,}$ Computation has been performed with Matlab, on a computer with an intel core i7 processor, a frequency of 2.7 GHz, and a RAM size of 16 GB.

provided in Fig. 2 and Fig. 3 between the actual and estimated sources obtained by different methods. Indeed, the actual sources used in Fig. 2 correspond to an example of Experiment 1 where N = 1000 and the random mixing matrix $\mathbf{A_{Exp1}}$ is given by

$$\mathbf{A_{Exp1}} = \begin{pmatrix} 0.0075 & 0.1461 & 0.3237\\ 0.1221 & 0.4249 & 0.0723\\ 0.4813 & 0.2845 & 0.5262 \end{pmatrix},$$
(32)

while Fig. 3 shows an example of Experiment 2 where the random mixing matrix A_{Exp2} is as follows

$$\mathbf{A_{Exp2}} = \begin{pmatrix} 0.1549 & 0.1405 & 0.3916\\ 0.5258 & 0.2041 & 0.9370\\ 0.2047 & 0.5108 & 0.4310 \end{pmatrix}.$$
 (33)

We notice in Fig. 2, which only shows 50 samples corresponding to $n \in [200, 249]$ for the sake of clarity, that both synthetic sources are much better estimated by our method than by the NMF-Grd-LQ method. Fig. 3 shows that both chemical sources are well estimated by our method, while only one source is well estimated by the NMF-Grd-LQ method.

In the following, our goal is to evaluate the performance of the proposed method with only one, then four lags, when the observed signals are corrupted by noise. Moreover, comparisons with the NMF-Grd-LQ method are carried out.

In the last two experiments, we performed 100 Monte Carlo simulations and, at each simulation, we modified the source signals and the mixing parameters in case of Experiment 3, and only the mixing parameters in case of Experiment 4.

The mean of the SIR and NMSE of the sources and of all the mixing parameters obtained for these two experiments are presented in Table 2 and Table 3. For clarity, SIR_s representation was performed versus the SNR values as shown in Fig. 4 (Experiment 3) and Fig. 5 (Experiment 4).

Considering Experiment 3, it is noted that the SIR_s obtained by the B-SO-BSS method are acceptable down to SNR equal to 40 dB. By comparing these results with those obtained by NMF-Grd-LQ, we notice that for high SNR values, B-SO-BSS gives the best results, however for low ones, NMF-Grd-LQ seems more efficient. Indeed, it can be seen that the SIR_s obtained by NMF-Grd-LQ remain acceptable down to SNR = 30 dB.

In the case of Experiment 4, the results obtained by the B-SO-BSS method are acceptable down to SNR equal to 50 dB. On the contrary, for all considered SNR values, NMF-Grd-LQ fails to yield accurate enough estimates.

In that regard, as expected, the presence of noise in the observed signals decreases the separation performance of our method. Nevertheless, performance remains acceptable for relatively high SNR values.

SNR (dB) Algorithms		60	50	40	30
$\boxed{ \begin{array}{c} \textbf{B-SO-BSS} \\ \tau_i = 1 \end{array} }$	SIR_s (dB)	25.46	20.49	13.60	6.13
	$NMSE_s$	0.03	0.08	0.21	0.43
	SIR_a (dB)	17.80	14.35	8.99	7.15
	$NMSE_a$	0.11	0.21	0.51	0.62
B-SO-BSS $\tau_i = 1, 2, 3, 4$	SIR_s (dB)	25.22	20.48	13.66	6.47
	$NMSE_s$	0.03	0.08	0.21	0.42
	SIR_a (dB)	18.17	14.52	9.21	7.29
	$NMSE_a$	0.10	0.21	0.51	0.61
NMF-Grd- LQ	SIR_s (dB)	14.82	14.61	13.93	12.77
	NMSEs	0.06	0.06	0.06	0.07
	SIR_a (dB)	11.00	10.98	10.62	10.60
	$NMSE_a$	0.18	0.19	0.19	0.20

 $\label{eq:and_state} \textbf{Table 2} \hspace{0.1 cm} \text{Simulation results obtained by Experiment 3 using B-SO-BSS with one and four lags and NMF-Grd-LQ when mixtures are corrupted by noise.}$

 Table 3
 Simulation results obtained by Experiment 4 using B-SO-BSS with one and four lags and NMF-Grd-LQ when mixtures are corrupted by noise.

Algorithm	SNR (dB)	60	50	40	30
$\boxed{ \begin{array}{c} \textbf{B-SO-BSS} \\ \tau_i = 1 \end{array} }$	SIR_s (dB)	13.59	13.10	9.55	5.10
	$NMSE_s$	0.2881	0.33	0.45	0.61
	SIR_a (dB)	20.69	19.83	17.93	14.87
	$NMSE_a$	0.38	0.51	0.70	0.80
B-SO-BSS $\tau_i = 1, 2, 3, 4$	SIR_s (dB)	16.60	15.57	10.11	6.64
	$NMSE_s$	0.24	0.36	0.66	0.74
	SIR_a (dB)	20.72	20.20	16.66	14.47
	$NMSE_a$	0.34	0.42	0.73	0.56
NMF-Grd- LQ	SIR_s (dB)	7.54	5.89	3.02	2.01
	$NMSE_s$	0.51	0.54	0.66	0.74
	SIR_a (dB)	6.43	6.43	6.37	6.36
	$NMSE_a$	0.41	0.41	0.41	0.41

5 Conclusion

In this paper, we proposed a new and fast BSS method, called *Bilinear Second-Order Blind Source Separation* (B-SO-BSS), which is an extension of linear SOS methods, to separate sources mixed according to the bilinear model. First, we studied the statistical properties of the different *extended sources* when the *actual sources* are zero-mean and when they are not. Then, we presented the different steps performed in order to separate the *actual centred sources* and to estimate the

actual mixing parameters. Finally, we presented the experimental results, obtained by our proposed method. As a first step, we evaluated the method separation performance when applied to noiseless artificial mixtures of synthetic or chemical sources. We, therefore, clearly noticed the effectiveness of our method as compared to the NMF-Grd-LQ method. As a second step, we evaluated the method robustness to noise, and as expected, we noticed that the presence of noise in the generated mixtures decreases the effectiveness of our method. However, the performance remained acceptable for relatively high SNR values.

References

- Abed-Meraim K, Belouchiani A, Hua Y (1996) Blind identification of a linearquadratic mixture of independent components based on joint diagonalization procedure. In: Proceedings of the International Conference on Acoustics, Speech, and Signal Processing (ICASSP-96), Atlanta, GA, USA, pp 2718–2721
- Almeida MSC, Almeida LB (2012) Nonlinear separation of show-through image mixtures using a physical model trained with ICA. Signal Processing 92(4):872– 884
- Ando RA, Duarte LT, Jutten C, Attux R (2015) A blind source separation method for chemical sensor arrays based on a second order mixing model. In: 23rd European Signal Processing Conference (EUSIPCO-2015), Nice, France, pp 933–937
- Bedoya G (2006) Non-linear blind signal separation for chemical solid-state sensor arrays, Ph.D. thesis, Universitat Politecnica de Catalunya
- Belouchrani A, Abed-Meraim K, Cardoso JF, Moulines E (1997) A blind source separation technique using second-order statistics. IEEE Transactions on Signal Processing 45(2):434–444
- Castella M (2008) Inversion of polynomial systems and separation of nonlinear mixtures of finite-alphabet sources. IEEE Transactions on Signal Processing 56(8):3905–3917
- Comon P, Jutten C (2010) Handbook of Blind Source Separation. Independent Component Analysis and Applications. Academic Press, Oxford
- Deville Y (2016) Blind Source Separation and Blind Mixture Identification Methods. John Wiley, DOI 10.1002/047134608X.W8300
- Deville Y, Duarte LT (2015) An overview of blind source separation methods for linear-quadratic and post-nonlinear mixtures. In: Proceedings of the 12th International Conference on Latent Variable Analysis and Signal Separation (LVA/ICA 2015), Springer International Publishing Switzerland, LNCS 9237, Liberec, Czech Republic, pp 155–167
- Deville Y, Hosseini S (2007) Blind identification and separation methods for linearquadratic mixtures and/or linearly independent non-stationary signals. In: Proceedings of the International Symposium on Signal Processing and Its Applications (ISSPA-2007), Sharjah, United Arab Emirates, pp 1–4
- Deville Y, Hosseini S (2009) Recurrent networks for separating extractable target nonlinear mixtures. Part I: non-blind configurations. Signal Processing 89(4):378–393
- Duarte LT, Tomazeli L, Jutten C, Moussaoui S (2011) Bayesian source separation of linear and linear-quadratic mixtures using truncated priors. Journal of Signal Processing Systems 65(3):311–323

- Duarte LT, Romano JMT, Jutten C, Chumbimuni-Torres KY, Kubota LT (2014) Application of blind source separation methods to ion-selective electrode arrays in flow-injection analysis. IEEE Sensors Journal 14(7):2228–2229
- Eches O, Guillaume M (2014) A bilinear-bilinear nonnegative matrix factorization method for hyperspectral unmixing. IEEE Geoscience and Remote Sensing Letters 11(4):778–782
- Hosseini S, Deville Y (2013) Recurrent networks for separating extractable target nonlinear mixtures. Part II: Blind configurations. Signal Processing 93:671–683
- Jarboui L, Hosseini S, Deville Y, Guidara R, Ben-Hamida A (2014) A new unsupervised method for hyperspectral image unmixing using a linear-quadratic model.
 In: Proceedings of the 1st International Conference on Advanced Technologies for Signal and Image Processing (ATSIP-2014), pp 423–428
- Jarboui L, Hosseini S, Guidara R, Deville Y, Ben-Hamida A (2016) A MAPbased NMF approach to hyperspectral image unmixing using a linear-quadratic mixture model. In: Proceedings of the International Conference on Acoustics, Speech and Signal Processing (ICASSP-2016), pp 3356–3360
- Krob M, Benidir M (1993) Blind identification of a linear-quadratic model using higher-order statistics. In: Proceedings of the International Conference on Acoustics, Speech, and Signal Processing (ICASSP-93), vol 4, pp 440–443
- Liu Q, Wang W (2013) Show-through removal for scanned images using non-linear NMF with adaptive smoothing. In: Proceedings of ChinaSIP, Beijing, pp 650–654
- Meganem I, Déliot P, Briottet X, Deville Y, Hosseini S (2014a) Linear-quadratic mixing model for reflectances in urban environments. IEEE Transactions on Geoscience and Remote Sensing 52(1):544–558
- Meganem I, Deville Y, Hosseini S, Déliot P, Briottet X (2014b) Linear-quadratic blind source separation using NMF to unmix urban hyperspectral images. IEEE Transactions on Signal Processing 62(7):1822–1833
- Merrikh-Bayat F, Babaie-Zadeh M, Jutten C (2011) Linear-quadratic blind source separating structure for removing show-through in scanned documents. International Journal on Document Analysis and Recognition (IJDAR) 14(4):319–333
- Taleb A (2002) A generic framework for blind source separation in structured nonlinear models. IEEE Transactions on Signal Processing 50:1819–1830
- Tong L, Soon VC, Huang YF, Liu R (1990) Amuse: a new blind identification algorithm. In: Proceedings of the International Symposium on Circuits and Systems (ISCAS-1990), vol 3, pp 1784–1787

Lina Jarboui received the Engineering degree in Computer Science in 2010 and the M.Sc. degree in New Technologies of Dedicated Computer Systems in 2012, both from the National Engineering School of Sfax (Tunisia). She is currently a Ph.D. student at the National Engineering School of Sfax (Tunisia) and the University of Toulouse (France), and a teaching and research assistant in digital image processing and scientific computing at the Conservatoire National des Arts et Métiers (France). Her main research concern Blind Source Separation methods, Signal and Image Processing, and Spectral Unmixing.

Yannick Deville was born in Lyon, France, in 1964. He graduated from the Ecole Nationale Supérieure des Télécommunications de Bretagne (Brest,

France) in 1986. He received the D.E.A and Ph.D. degrees, both in Microelectronics, from the University of Grenoble (France), in 1986 and 1989 respectively. From 1986 to 1997, he was a Research Scientist at Philips Research Labs (Limeil, France). His investigations during that period concerned various fields, including GaAs integrated microwave RC active filters, VLSI cache memory architectures and replacement algorithms, neural network algorithms and applications, and nonlinear systems. Since 1997, he has been a Professor at the University of Toulouse (France). From 1997 to 2004, he was with the Acoustics lab of that University. Since 2004, he has been with the Astrophysics lab in Toulouse, which is part of the University and of the French National Center for Scientific Research (CNRS). His current major research interests include signal and image processing (especially hyperspectral data), higher-order statistics, time-frequency analysis, neural networks, quantum entanglement phenomena, and especially Blind Source Separation and Blind Identification methods (including Independent or Sparse Component Analysis and Nonnegative Matrix Factorization) and their applications to Remote Sensing, Astrophysics and Quantum Information Processing.

Shahram Hosseini was born in Shiraz, Iran, in 1968. He received the B.Sc. and M.Sc. degrees in electrical engineering from the Sharif University of Technology, Tehran, Iran, in 1991 and 1993, respectively, and the Ph.D. degree in signal processing from the Institut National Polytechnique, Grenoble, France, in 2000. He is currently an Associate Professor at the Université Paul Sabatier Toulouse 3, Toulouse, France. His research interests include blind source separation, artificial neural networks, and adaptive signal processing.

Rima Guidara was born in Sfax, Tunisia, in 1982. In 2005, she graduated as an electronics and signal processing engineer from the Ecole Nationale Supérieure d'Electrotechnique, d'Electronique, d'Informatique, d'Hydraulique et des Télécommunications (ENSEEIHT), Toulouse, France, and received the M.Sc. degree in image and signal processing from the National Polytechnic Institute of Toulouse. In 2009, she received the Ph.D. degree in signal and image processing from the University of Toulouse 3, France. Since 2009, she has been with the Ecole Nationale dElectronique et des Télécommunications (ENETCom) in Sfax, Tunisia, where she is currently an assistant professor. Her main research field concerns signal and image processing and their applications to remote sensing. Her current major research interests include blind source separation, multi- and hyperspectral data unmixing and classification, multisensor data fusion, LIDAR point cloud processing and 3D segmentation, canopy structure and biomass estimation.

Ahmed Ben Hamida born in 1963 in Sfax (Tunisia), graduated in electrical engineering in 1988, received his masters degree in telecommunications from ENIT School of Engineering, Tunis, Tunisia, in 1993, Doctorate degree in 1998, and HDR degree for research director from ENIS School of Engineering, Sfax, Tunisia, in 2003. Currently, he is a professor at the ENIS National School of Engineering, Sfax, Tunisia, and he is the head of the research unit Advanced Technologies for Medicine and Signals 'ATMS'. Undertaken research was mainly directed according to speech processing and medical image processing with all modalities.

Leonardo Tomazeli Duarte received the B.S. and M.Sc. degrees in electrical engineering from the University of Campinas (UNICAMP), Brazil, in 2004 and 2006, respectively, and the Ph.D. degree from the Grenoble Institute of Technology (Grenoble INP), France, in 2009. Since 2011, he has been with the School of Applied Sciences (FCA) at UNICAMP, Limeira, Brazil, where he is currently an assistant professor. His research interests are mainly associated with the theory of unsupervised signal processing and include signal separation, independent component analysis, nonlinear models, Bayesian methods, and applications in chemical sensors, quality control and seismic signal processing. He is a Senior Member of the IEEE.



Fig. 2 Comparison between estimated and actual synthetic sources, in the case of a noiseless artificial BL mixture, using 3 different methods: B-SO-BSS ($\tau_i = 1$) (top), B-SO-BSS ($\tau_i = 1, 2, 3, 4$) (middle) and NMF-Grd-LQ (bottom). Only a part of signals corresponding to $n \in [200, 249]$ is shown here.

Samples

Samples



Fig. 3 Comparison between estimated and actual chemical sources, in the case of a noiseless artificial BL mixture, using 3 different methods: B-SO-BSS ($\tau_i = 1$) (top), B-SO-BSS ($\tau_i = 1, 2, 3, 4$) (middle) and NMF-Grd-LQ (bottom).



Fig. 4 SIR_s versus SNR in the case of Experiment 3.



Fig. 5 SIR_s versus SNR in the case of Experiment 4.

23